Table of Contents:

Introduction

General

High Frequency Case

Low Frequency Case

Introduction:

General: pg. 2

From Gardiol, Lossy Transmission Lines, p. 168:

$$\alpha + \beta \cdot \mathbf{j} = \sqrt{(G' + \omega \cdot C' \cdot \mathbf{j}) \cdot (R' + \omega \cdot L' \cdot \mathbf{j})}$$

where: α = attenuation in nepers/U.L. (can be derived from measurements)

 β = phase shift in radians/U.L. (can be derived from measurements)

ω= frequency in radians/second (known)

G'= conductance in Siemans/U.L. (unknown)

C'= capacitance in Farads/U.L. (usually given by cable manufacturer or can easily be measured using a short length of line)

R'= resistance in Ohms/U.L. (unknown)

L'= inductance in Henrys/U.L. (unknown)

U.L.= Unit Length

which can be rewritten as: $(\alpha + \beta \cdot j)^2 = (G' + \omega \cdot C' \cdot j) \cdot (R' + \omega \cdot L' \cdot j)$

expanding the expression, yields:

$$\alpha^2 + 2 \cdot \alpha \cdot \beta \cdot j - \beta^2 = R' \cdot G' + \omega \cdot G' \cdot L' \cdot j + \omega \cdot R' \cdot C' \cdot j - \omega^2 \cdot C' \cdot L'$$

collecting real and imaginary terms yields two equations:

$$\alpha^2 - \beta^2 = R' \cdot G' - \omega^2 \cdot C' \cdot L'$$
 and $2 \cdot \alpha \cdot \beta = \omega \cdot G' \cdot L' + \omega \cdot R' \cdot C'$

solving for α in the second yields: $\alpha = \frac{\omega \cdot G' \cdot L'}{2 \cdot \beta} + \frac{\omega \cdot R' \cdot C'}{2 \cdot \beta}$

However, we have three unknowns. We need three equations.

High Frequency Case:

pg. 3

The third equation can be obtained by making measurements at two different frequencies to separate the conductor and dielectric losses.

Let
$$A_1 = m \cdot \sqrt{f_1} + n \cdot f_1$$
 and $A_2 = m \cdot \sqrt{f_2} + n \cdot f_2$

where: A_1 = measured attenuation in decibels/U.L. at frequencf₁

 A_2 = measured attenuation in decibels/U.L. at frequencf₂

m,n = constants

U.L.= unit length

solving for m and n yields:

$$m = \frac{A_1 \cdot f_2 - A_2 \cdot f_1}{f_2 \cdot \sqrt{f_1 - f_1 \cdot \sqrt{f_2}}} \qquad n = \frac{A_2 \cdot \sqrt{f_1 - A_1 \cdot \sqrt{f_2}}}{f_2 \cdot \sqrt{f_1 - f_1 \cdot \sqrt{f_2}}}$$

These equations can now be used to solve for the total attenuation, A, at any frequency, f, as well as the conductor loss, Ac, and the dielectric loss, Ad.

$$A=Ac + Ad$$
 $Ac=m\cdot\sqrt{f}$

$$Ad=n\cdot f$$

in dB/U.L.

$$\alpha = \frac{A}{\left(\frac{20}{\ln(10)}\right)}$$

$$\alpha c = \frac{Ac}{\left(\frac{20}{\ln(10)}\right)}$$

$$\alpha = \frac{A}{\left(\frac{20}{\ln(10)}\right)} \qquad \alpha c = \frac{Ac}{\left(\frac{20}{\ln(10)}\right)} \qquad \alpha d = \frac{Ad}{\left(\frac{20}{\ln(10)}\right)} \qquad \text{in Nepers/U.L.}$$

β can be measured or derived from cable maufacturer's data by the following formula:

$$\beta = \frac{\omega}{vp}$$
 where: $vp = relative phase velocity (dimensionless)$

Now we have three equations to find L', R', and G' since,

$$\alpha c = \frac{\omega \cdot G' \cdot L'}{2 \cdot \beta} \qquad \alpha d = \frac{\omega \cdot R' \cdot C'}{2 \cdot \beta}$$

and previously derived,
$$\alpha^2 - \beta^2 = R' \cdot G' - \omega^2 \cdot C' \cdot L'$$

Solving for L', R', and G' yields: $L' = \frac{\left(\beta^2 - \alpha^2\right) + \sqrt{\left(\beta^2 - \alpha^2\right)^2 + 16 \cdot \beta^2 \cdot \alpha c \cdot \alpha d}}{2 \cdot \omega \cdot C'}$

and:
$$R' = \frac{2 \cdot \beta \cdot \alpha c}{\omega \cdot C'}$$

$$G' = \frac{2 \cdot \beta \cdot \alpha d}{\omega \cdot L'}$$

Low Frequency Case:

At low frequencies, we can usually assume that the dielectric loss is negligible.

Therefore $\alpha d=0$, G'=0, and $\alpha=\alpha c$.

This results in two equations:
$$L' = \frac{\beta^2 - \alpha^2}{\omega \cdot C'}$$

$$R' = \frac{2 \cdot \beta \cdot \alpha}{\omega \cdot C'}$$

However at low frequencies, the attenuation no longer varies as the square root of the frequency since the skin depth δ approaches the radius or thickness of the smallest conductor. Also the characteristic impedance of the line can change dramatically with frequency, so that α may no longer be measured into fixed and matched source and load impedances such as 50 ohms. If the loss, A, in decibels/U.L. is measured into a matched (at high frequencies) source and load impedance such as 50 or 75 ohms real, then the attenuation into matched source and load impedances, α , must be found first by iteration using the formula for the characteristic impedance of the line along with the formula for the loss into given source and load impedances.

Similarly β no longer varies linearly with frequency.....

From Gardiol, p. 200:

$$\frac{\frac{P_{L}}{P_{M}}}{=} \frac{4 \cdot R_{G} \cdot R_{L} \cdot (\left|Y_{c}\right|)^{2}}{\left[\left|\left(Z_{G} \cdot Z_{L} \cdot Y_{c}^{2} + 1\right) \cdot \sinh(\gamma \cdot L) - \left(Z_{G} + Z_{L}\right) \cdot Y_{c} \cdot \cosh(\gamma \cdot L)\right|\right]^{2}}$$

$$Y_{c} = \sqrt{\frac{G' + \omega \cdot C' \cdot j}{R' + \omega \cdot L' \cdot j}} \qquad \gamma = \alpha + \beta \cdot j \qquad A = 10 \cdot log \left(\frac{P_{L}}{P_{M}}\right)$$

Start the iteration with:
$$\alpha = \frac{A}{\left(\frac{20}{\ln(10)}\right)}$$

M = READPRN (LDF5_50A)

 $f = M^{<0>}$

A = M >

vp = .89

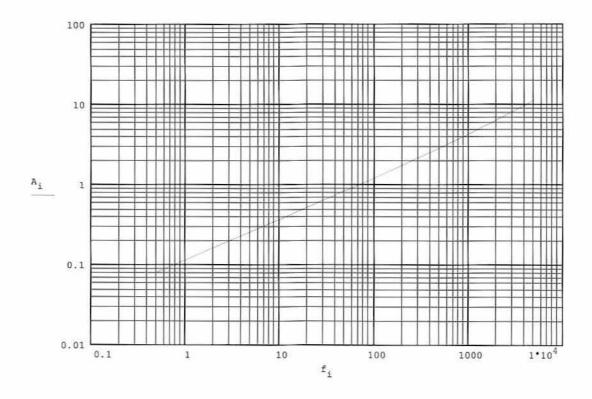
Andrew LDF5-50A.MCD manufacturer's data

A

T	0.5	0.0804
	1	0.115
	1.5	0.141
	2	0.164
	10	0.367
M =	20	0.525
	30	0.646
	50	0.843
	88	1.13
	100	1.21
	108	1.26
	150	1.5
	174	1.63
	200	1.76
	300	2.19
	400	2.56
	450	2.74
	500	2.9
	512	2.94
	600	3.21
	700	3.5
	800	3.78
	824	3.85
	894	4.03
	960	4.2
	1·10 ³	4.3
	1.25·10 ³	4.9
	1.5.103	5.45
	1.7.103	5.87
	2 • 10 ³	6.46
	2.3.103	7.05
	3 • 10 ³	8.31
	4 • 10 ³	9.94
	5 • 10 ³	11.5
L	5 • 10 ³	11.5

N = 33

i = 0..N



i = 0..N - 1

$$\mathbf{m_i} = \frac{\mathbf{A_i \cdot f_{i+1}} - \mathbf{A_{i+1} \cdot f_{i}}}{\mathbf{f_{i+1} \cdot \sqrt{f_{i}}} - \mathbf{f_i \cdot \sqrt{f_{i+1}}}}$$

$$n_{i} = \frac{A_{i+1} \cdot \sqrt{f_{i}} - A_{i} \cdot \sqrt{f_{i+1}}}{f_{i+1} \cdot \sqrt{f_{i}} - f_{i} \cdot \sqrt{f_{i+1}}}$$

 $\mathbf{m}_{\mathbf{N}} := \mathbf{m}_{\mathbf{N}-1}$

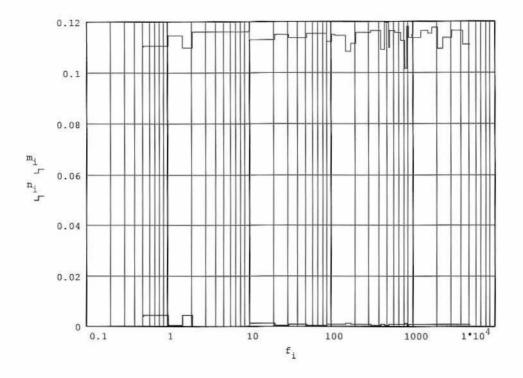
$$n_{N} := n_{N-1}$$

i = 0..N

fi
0.5
1
1.5
2
10
20
30
50
88
100
108
150
174
200
300
400
450
500
512
600
700
800
824
894
960
1.103
1.25.103
1.5.103
1.7.103
2·10 ³
2.3·10 ³
3·10 ³
4·10 ³
5·10 ³

mi	
0.	1106
0.	1144
0.	1097
0.	1159
0.	1128
0.	1149
0.	1136
0.	1154
0.	1123
0.	1148
0.	1143
0.	1083
0.	1114
0.	1156
0.	1164
0.	1088
0.	1194
0.	1097
0.	1164
0.	1156
0.	1127
0.	1015
0.	1182
0.	1135
0	.115
0.	1138
0.	1163
0.	1152
0.	1178
0.	1092
0.	1138
0.	1165
0.	1108
0.	1108

n _i	
0.004	
5.6072.10) 4
0.004	
5.153.10) - 5
0.001	_
5.4657.10	
8.0013.10	
5.3694.10	
8.7471.10	
6.2083.10	
6.6352.10)-4
0.001	41-1
9.2597.10	
6.2577.10) - 4
5.8231.10	
9.6013-10)-4
4.5936.10) - 4
8.9574.10) - 4
5.9804.10) - 4
6.3174.10)-4
7.4209-10) - 4
0.001	
5.5445.10	
7.1128.10	
6.6288.10	
7.0059.10	
6.2986.10) - 4
6.5968.10) - 4
5.9639.10) - 4
7.886.10)- 4
6.9218.10) - 4
6.4273.10) - 4
7.3265.10) - 4
7.3265.10) - 4
	_



$$Ac_i = m_i \cdot \sqrt{f_i}$$
 $Ad_i = n_i \cdot f_i$

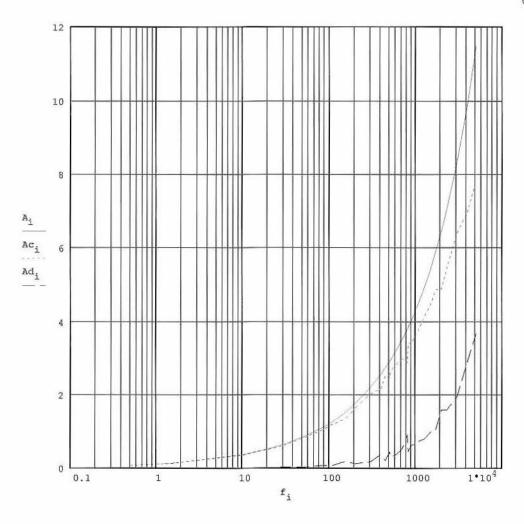
$$Ad_i = n_i \cdot f_i$$

0.5		
1		
1.5		
2		
10		
20		
30		
50		
88		
100		
108		
150		
174 200		
300		
400	_	
450		
500	_	
512		
600		
700		
800		
824		
894		
960		
1·10 ³		
1.25.	10	
1.5.	103	3
1.7.	10	3
55 CEMED 12	10	3
	103	3
		3
-12 A	10	
	TO	
5.	10	5

0.	0782
0.	1144
0.	1344
0.	1639
0.	3568
0.	5141
0	.622
0.	8162
1	.053
1.	1479
1.	1883
1.	3258
1.	4689
1.	6348
2.	0153
2.	1759
2.	5333
2.	4521
2.	6338
2	.831
2.	9805
2.	8721
3.	3931
3.	3941
3.	5636
3.	5994
4.	1127
4.	4605
4.	8561
4.	8828
5	.458
6.	3818
7.	0094

Ad _i	0.0022
E 6	072-10
5.0	0.0066
	GIS ALTONOMIC COLOR
1.0	306-10-4
	0.0102
	0.0109
	0.024
	0.0268
	0.077
	0.0621
	0.0717
	0.1742
	0.1611
	0.1252
	0.1747
	0.3841
	0.2067
	0.4479
	0.3062
	0.379
	0.5195
	0.9079
	0.4569
	0.6359
	0.6364
	0.7006
	0.7873
	0.9895
	1.0139
	1.5772
	1.592
	1.9282
	2.9306
	3.6633

A	i.
(0.0804
	0.115
	0.141
	0.164
	0.367
	0.525
	0.646
1	0.843
Ī	1.13
-	1.21
	1.21 1.26 1.5
	1.63
ĺ	1.76
	2.19
	2.56
Ī	2.74
	2.56 2.74 2.9
	2.94
Ī	3.21
Ī	3.5
	3.78
Ī	3.85
	4.03
	4.2
	4.3
	4.9
Ī	5.45
	5.87
	6.46
Ī	7.05
	8.31
	9.94
Ī	11.5



Let:

$$\alpha_{\mathbf{i}} := \frac{A_{\mathbf{i}}}{\left(\frac{20}{\ln{(10)}}\right) \cdot 100} \qquad \alpha_{\mathbf{c}_{\mathbf{i}}} := \frac{Ac_{\mathbf{i}}}{\left(\frac{20}{\ln{(10)}}\right) \cdot 100} \qquad \alpha_{\mathbf{d}_{\mathbf{i}}} := \frac{Ad_{\mathbf{i}}}{\left(\frac{20}{\ln{(10)}}\right) \cdot 100}$$

$$c := 2.9979 \cdot 10^{8} \qquad f_{26} = 1.25 \cdot 10^{3} \qquad \alpha_{26} = 0.0056 \qquad \frac{20}{\ln{(10)}} = 8.6859$$

$$vp := .89 \qquad \omega_{\mathbf{i}} := 2 \cdot \pi \cdot f_{\mathbf{i}} \cdot 10^{6} \qquad \beta_{\mathbf{i}} := \frac{\omega_{\mathbf{i}}}{vp \cdot c} \qquad C := 75.0 \cdot 10^{-12}$$

$$\omega_{26} = 7.854 \cdot 10^{9} \qquad \beta_{26} = 29.4363 \qquad A_{26} = 4.9$$

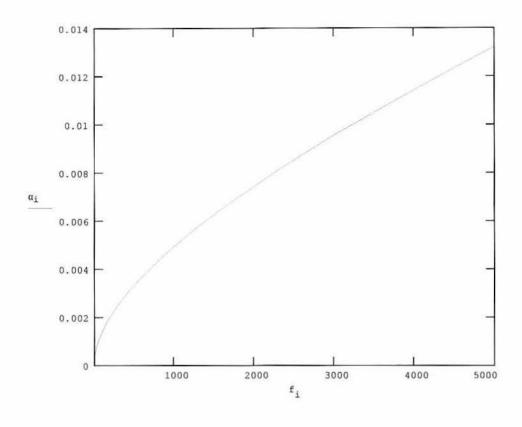
$$\mathbf{L_{i}} := \frac{\left[\ \left(\boldsymbol{\beta_{i}} \right)^{2} - \left(\boldsymbol{\alpha_{i}} \right)^{2} \ \right] + \sqrt{\left[\ \left(\boldsymbol{\beta_{i}} \right)^{2} - \left(\boldsymbol{\alpha_{i}} \right)^{2} \ \right]^{2} + 16 \cdot \left(\boldsymbol{\beta_{i}} \right)^{2} \cdot \alpha \mathbf{c_{i}} \cdot \alpha \mathbf{d_{i}}}{2 \cdot \left(\boldsymbol{\omega_{i}} \right)^{2} \cdot C}$$

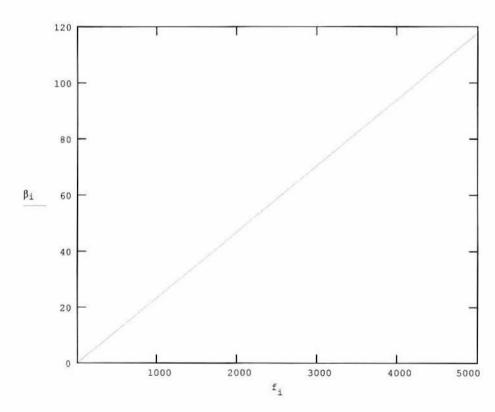
$$R_{i} = \frac{2 \cdot \beta_{i} \cdot \alpha c_{i}}{\omega_{i} \cdot C} \qquad G_{i} = \frac{2 \cdot \beta_{i} \cdot \alpha d_{i}}{\omega_{i} \cdot L_{i}}$$

f _i	3				
1	_	_	_	_	_
1.5	e e	_			
2					
10					
20				Ī	
30					
50					
88					
100					
108					
150	Ö.				
174	ľ				
200					
300					
400					
450		_			
500 512			_		
512 600					
700		_			_
800				_	_
824	100	_	_	_	_
894	H	_	_		
960		_	_	_	-
1.1		3			
1 0	5	_	1	0	3
Carlo Carlo		_		1000	3
1.	5	•	_	0	3
1.	7	٠	-	0	
	2	•	1	0	
2.	3		1	0	3
	3		1	0	3
	4	•	1	0	3
	5	-	_	0	3

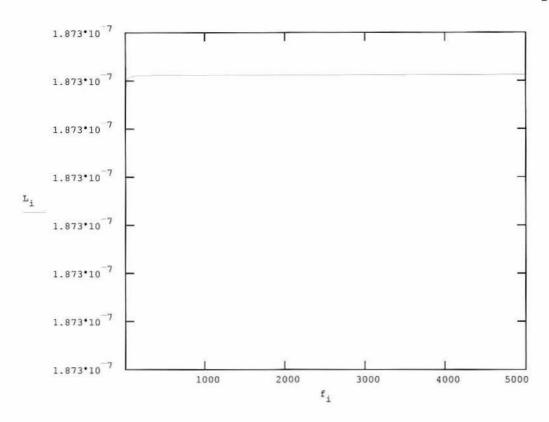
$\alpha_{\mathbf{i}}$	
9.2564.10	5
1.324.10	4
1.6233.10	4
1.8881.10	4
4.2252.10	4
6.0443-10	4
7.4373.10	4
9.7054-10	4
0.0013	ī
0.0014	
0.0015	
0.0017	
0.0019	
0.002	
0.0025	
0.0029	
0.0032	
0.0033	
0.0034	
0.0037	
0.004	
0.0044	
0.0044	
0.0046	
0.0048	
0.005	
0.0056	
0.0063	
0.0068	
0.0074	
0.0081	
0.0096	
0.0114	
0.0132	
	_

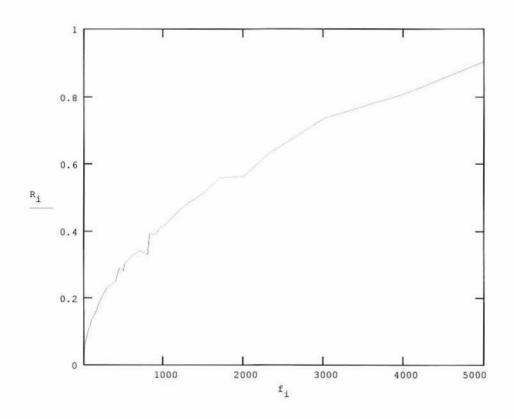
3	
C	
0	
C	
C	
-1	0.471
C	
1	
2	
2	
2	
7	
4	.0975
4	
7	.0647
9	.4196
1	0.5971
1	1.7745
1	2.0571
1	4.1294
1	6.4843
	8.8392
	9.4044
	1.0528
	2.6071
	23.549
2	9.4363
	5.3235
	0.0333
	47.098
	4.1627
	70.647
	4.1961
100	4.1301



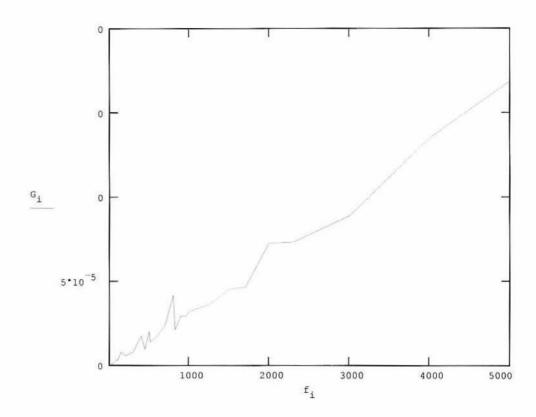


f _i	L _i	$R_{\underline{i}}$	$G_{\mathtt{i}}$
0.5	1.8728 • 10 - 7	0.009	1.0204-10 ⁻⁷
1	1.8729·10 ⁻⁷	0.0132	2.5837.10-8
1.5		0.0155	3.0624 • 10 - 7
2	1.8729 • 10 - 7	0.0189	
10	1.8729.10-7	0.0411	4.7488.10-9
20	1.8729-10 ⁻⁷	0.0592	4.7066.10-7
30	1.8729 • 10 - 7	0.0716	5.0369.10
50 88	1.8729.10-7	0.0939	1.106.10-6
100	1.8729.10-7	0.1321	1.237.10-6
108	1.8729 • 10 - 7	0.1367	3.5468.10-6
150	1.8729·10 ⁻⁷	0.1526	2.8606.10
174		0.169	
200	1.8729·10 ⁻⁷	0.1881	3.3019-10-6
300	1.8729.10-7	0.2319	8.0253.10-6
400	1.8729.10-7	0.2504	7.4239.10-6
450	1.8729.10-7	0.2915	5.7667·10 ⁻⁶
500	1.8729.10-7	0.2822	8.0493-10-6
512	1.8729 • 10 - 7	0.3031	1.7696.10-5
700	1.8729-10-7	0.3257	9.5247.10-6
800	1.8729 • 10 - 7	0.345	2.0637-10-5
824	1.8729·10 ⁻⁷	0.3904	1.4109-10-5
894		0.3905	
960	1.8729·10 ⁻⁷	0.4101	1.7465.10-5
1·10 ³	1.8729 • 10 - 7	0.4142	2.3935.10-5
1.25.103	1.8729.10-7	0.4732	4.1833.10-5
1.5.103	1.8729 • 10 - 7	0.5132	2.1051.10 ⁻⁵
	1.8729.10-7	0.5588	2.93.10-5
1.7.103	1.8729-10-7	0.5618	2.9322.10-5
2·10 ³	1.8729·10 ⁻⁷	0.628	3.2281.10-5
2.3.103	1.8729-10-7	0.8065	3.6278-10-5
3·10 ³	1.8729·10 ⁻⁷	0.9017	4.5594.10
4·10 ³			
5·10 ³	1.8729 • 10 - 7		4.6716.10
	1.8729 • 10 - 7		7.2673.10
	1.8729.10 ⁻⁷		7.3355.10-5
	1.8729.10 ⁻⁷		8.8845.10-5
	1.8729.10-7		1.3503.10-4
	1.8729 • 10 - 7		1.6879-10-4
			A STATE OF THE PARTY OF THE PAR





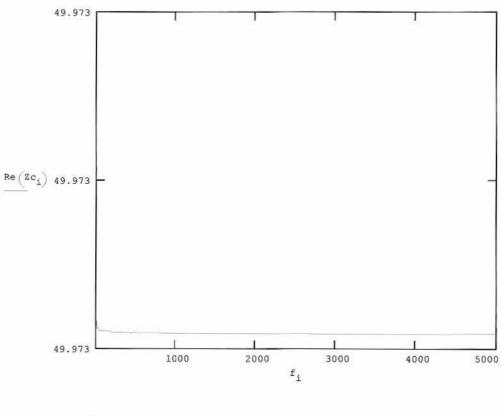
e i

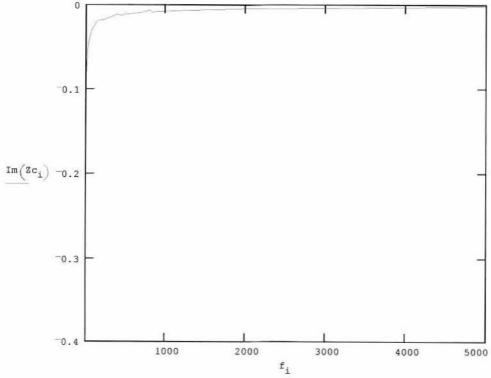


$$Zc_{i} = \sqrt{\frac{R_{i} + \omega_{i} \cdot L_{i} \cdot j}{G_{i} + \omega_{i} \cdot C \cdot j}}$$

fi
0.5
1
1.5
2
10
20
30
50
88
100
108
150
174
200
300
400
450
500
512
600
700
800
824
894
960
1.103
1.25.103
1.5·10 ³
1.7·10 ³
2·10 ³
2.3.103
3·10 ³
4·10 ³
5·10 ³

49.9727		0.3712
49.9726		0.2782
49.9726	_	0.208i
49.9726	-	0.2001
49.9726	-	0.0847
49.9726		0.0615
49.9726	-	0.0487
49.9726		0.0386
49.9726	-	0.0271
49.9726	_	0.0265
49.9726	-	0.0253
49.9726	2	0.0188
49.9726	-	0.0184
49.9726	_	0.0184
49.9726	-	0.015j
49.9726	_	0.0109
49.9726	#	0.0126
49.9726	_	0.0098
49.9726	_	0.0111
49.9726	-	0.01j
49.9726	-	0.0086
49.9726	7	0.006j
49.9726	-	0.0087
49.9726	-	0.0075
49.9726	-	0.0074
49.9726		0.0071
49.9726	-	0.0065
49.9726		0.0057
49.9726	**	0.0055
49.9726	e.	0.004j
49.9726	П	0.0041
49.9726	ė	0.0036
49.9726	Ų.	0.0025
49.9726	-	0.002j







Superior to Braided Cable

Solid copper corrugated outer conductor results in low loss, high power handling and continuous RFI/EMI shielding to minimize interference and maximize system security. Cable can be formed to a 10 in (250 mm) radius.

Weatherproof

Annular corrugations prevent water migration. Connector O-rings seal out moisture. Closed cell foam prevents water penetration.

Quick and Easy Connector Attachment

Patented, self-flaring design.

Low Loss Foam

Pressurization not required.

Proven performance in applications such as:

- Industry standard in land mobile radio.
- Cellular radio.
- Phase stabilized versions for phased array radars.
- VLF and HF communications systems; AM and FM radio broadcast.
- Mil-spec versions available.

Low VSWR Cable

Type LDF5P-50A is a low-VSWR version of LDF5-50A. Low VSWR specifications are tabulated on the right. Achievable VSWR is a function of maximum operating frequency, connector interfaces and cable length. The figures are guaranteed for factory assemblies and are typical for field cut lengths. If two different interfaces are used, the higher VSWR value is the guarantee.

Standard operating frequency bands include those commonly used for terrestrial microwave and satellite communication earth station applications:

Terrestrial Microwave Low VSWR specifications

for frequency bands (Specify bands):

1.427-1.535 GHz 1.15 (23.1) 2.11-2.2 GHz 1.12 (24.9) 1.7-1.9 GHz 1.15 (23.1) 1.7-2.11 GHz 1.15 (23.1) 1.85-1 99 GHz 1.12 (24.9) 1.9-2.3 GHz 1.15 (23.1) 1.99-2.11 GHz 1.15 (23.1) 2.3-2.7 GHz 1.20 (20.8)

Low VSWR cable for cellular radio is listed in the

"Characteristics" table.

Earth Station 3.625-4 2 GHz

7/8" Foam Dielectric

7/8"



Characteristics

Nominal Size

Impedance, ohms	50***			
Cable Type Numbers				
Standard Cable, Standard Jacket	LDF5-50A*			
Standard Cable, Fire-Retardant,	LDF5RN-50A			
Non-Halogenated Jacket				
Specially Tested and Selected Cable				
Low-VSWR Cable	LDF5P-50A			
(Specify Operating Band, see table below)				
Cellular Radio				
824-894 MHz, 1.20 max. VSWR	42150B-48			
880-960 MHz, 1.20 max. VSWR	42150B-54			
Qualified to MIL-C-28830/4	202071-2			

Electrical Characteristics

Maximum Frequency, GHz	5.0
Velocity, percent	89
Peak Power Rating, kW	44
DC Resistance, ohms/1000 ft (1000 m)	
Inner	0.35 (1.15)
Outer	0.36 (1.18)
DC Breakdown, volts	6000
Jacket Spark, volts RMS	8000
Capacitance, pF/ft (m)	22.8 (75.0)
Inductance, µH/ft (m)	0.057 (0.187)

Mechanical Characteristics

A SHALL STORY OF THE STORY OF T
Copper
Copper
m) 1.09 (28)
r Conductor, in (mm) 0.98 (24.9)
n) 12 (16.3)
0.33 (0.49)
325 (147)
b/in (kg/mm) 80 (1.4)
n (typical) 15 n) 12 (0.33 (325

^{*}For broadcast applications, specify TV channel or frequency.

Low VSWR Specifications, Type LDF5P-50A

Up to	Using Connector	Assembly VSWR, Maximum (R.L., dB					
Freq. GHz	Type No.	to 25 ft (8 m)	25 - 100 ft (8 - 30 m)	100 - 200 ft (30 - 60 m)			
1.7†	L45W (N Plug)	1.10 (26.4)	1.20 (20.8)	1.30 (17.7)			
	L45N (N Jack)	1.12 (24.9)	1.22 (20.1)	1.33 (17.0)			
	L45F or L45R	1.10 (26.4)	1.20 (20.8)	1.30 (17.7)			
2.7†	L45W (N Plug)	1.10 (26.4)	1.20 (20.8)	1.30 (17.7)			
	L45N (N Jack)	1.15 (23.1)	1.25 (19.1)	1.35 (16.6)			
	L45F or L45R	1.15 (23.1)	1.25 (19.1)	1.35 (16.6)			
4.2	L45W (N Plug)	1.10 (26.4)	1.20 (20.8)	1.35 (16.6)			
5.0	L45W (N Plug)	1.15 (23.)	1.20 (20.8)	1.35 (16.6)			

†See "Terrestrial Microwave" on the left for data on specific narrow bands.

^{***}A 75-ohm 7/8" diameter cable is available. Contact Andrew for further information.

Attenuation and Average Power

Frequency MHz	Attenuation dB/100 ft	Attenuation dB/100m	Average Power kW	Frequency MHz	Attenuation dB/100 ft	Attenuation dB/100 m	Average Power kW
0.5	0.0245	0.0804	44.0	500	0.885	2 90	2.25
1	0.0350	0.115	44.0	512	0.896	2.94	2.22
1.5	0.0431	0.141	44.0	600	0.979	3 21	2.03
2	0.0500	0.164	40.0	700	1.07	3 50	1.86
10	0.112	0.367	17.7	800	1,15	3.78	1.73
20	0.160	0.525	12.4	824	1.17	3.85	1.70
30	0.197	0.646	10.1	894	1.23	4.03	1.62
50	0.257	0.843	7.74	960	1.28	4.20	1 56
88	0.345	1.13	5.75	1000	1.31	4.30	1.52
100	0.369	1.21	5.38	1250	1.49	4.90	1 33
108	0.384	1.26	5.17	1500	1.66	5.45	1.20
150	0.458	1.50	4.34	1700	1.79	5.87	1.11
174	0.496	1.63	4.01	2000	1.97	6.46	1.01
200	0.535	1.76	3.72	2300	2.15	7.05	0.926
300	0.668	2.19	2.98	3000	2.53	8.31	0.785
400	0.781	2.56	2.55	4000	3.03	9.94	0.656
450	0.834	2.74	2.39	5000	3.50	11.5	0.568

Standard Conditions:

For Attenuation. VSWR 1.0, ambient temperature 24°C (75°F).

For Average Power, VSWR 1.0, ambient temperature 40°C (104°F), inner conductor temperature 100°C (212°F).

Connectors

Interface - See photos on pages 332 and 333	Type No.	Length in (mm)	Body Dia. in (mm)	Flange Dia. in (mm)	Weight Ib (kg)
"F" (male) connects with "F"-Series antennas	L45F	1.76 (44.7)	1.40 (35.6)	2.25 (57.2)	1.5 (0.7)
"F" Flange (female) for connection to jumper cable	48041	1.76 (44.7)	1.40 (35.6)	2.25 (57.2)	1.5 (0.7)
7/8 EIA Flange, no gas barrier at interface, includes inner connector	L45R	3.32 (84.3)	1.35 (34 3)	2.25 (57.2)	1.5 (0.7)
7/8 EIA Flange, right angle, no gas barrier at interface, includes inner connector	124800-1	3.94 (100.0)	1.34 (34.0)	2.25 (57 2)	1.5 (G.7)
N Plug (male), mates with UG-23	L45W	2.83 (71.9)	1.37 (34.8)		1.5 (0.7)
N Plug (male), low VSWR, mates with UG-23	L45EW*	2.83 (71.9)	1.37 (34.8)	7.5	1.5 (0.7)
N Jack (female), mates with UG-21	L45N	2.80 (71.1)	1.35 (34.3)	-	1.5 (0.7)
UHF Plug (male), mates with SO-239A	L45P	2.70 (68.5)	1.35 (34.3)	-	1.5 (0.7)
UHF Jack (female), mates with PL-259A	L45U	2.68 (68.1)	1.35 (34.3)	-	1.5 (0.7)
LC Plug (male), mates with UG-352	L45M	3.69 (93.7)	1.34 (34.0)		1.5 (0 7)
LC Jack (female), mates with UG-154	L45L	3.42 (86.8)	1.35 (34.3)		1.5 (0.7)
HN Plug (male), mates with UG-60	L45J	2.95 (74.9)	1.34 (34.0)	-	1.5 (0.7)
7/16 DIN male	L45DM	2.63 (66.7)	1.38 (35.1)	142	1.5 (0.7)
7/16 DIN female	L45DF	2.72 (69.1)	1.36 (34.5)	-	1.5 (0.7)
End Terminal, for strap connection to center conductor	L45T	4.88 (123.8)	1.35 (34.3)	*	1.5 (0.7)
Splice Connector Pin-Paks, five replacement pins	L45Z	3.34 (84.8)	1.47 (37.3)	2	1.5 (0.7)
For L45W	43158-5	_	-	1722	120
For L45N	43157-2		_	-	-

For RF connector adaptors, see page 334.

Accessories - See page 301

To Order

- A sample order is shown on page 273.
- Specify cable Type Number and length in feet or metres.
 See "Characteristics" table.
- For low-VSWR cable, specify the operating frequency band (see "Low-VSWR Cables" for standard frequency bands and VSWR/Return Loss specifications).
- Specify connector Type Numbers and "attached" or "unattached". When attached connectors on an assembly are different, specify which is "first off" the reel.

Further Information

For general information on HELIAX coaxial cables see pages 268-273.



^{*}Connector for low-VSWR applications. Includes gold-plated inner connector and nickel-plated body.